



Ampère
Unité Mixte de Recherche CNRS
Génie Électrique, Électromagnétisme, Automatique, Microbiologie environnementale et Applications

State of the art of high switching frequency inductive DC-DC converters in SoC context

B. Allard, Florian Neveu, C. Martin

Ampère, Lyon

www.ampere-lab.fr

- Supplying large system on chip requires specific grid and is spread over multiple voltage domains. Each domain depends on a dedicated DC-DC. For the sake of integration, switch capacitor DC-DC seems to have the favors of designers but further increase in the switching frequency associated to a 3D approach gives back some competitiveness to inductive DC-DC converters.
- Some recent experiments in the +100 MHz switching frequency area demonstrate the interest of multiphase and coupled inductors. A global analysis of pertinent results as state of the art enables to draw similar conclusions and many other trends.
- Various landscapes have been populated with available figures in pertinent papers about implemented DC-DC converters. Trends are analyzed to exhibit design trade-offs and further extrapolations, focusing mainly on active devices.
- In the context of POWERSWIPE project, a 200 MHz DC-DC converter (3.3 V input voltage, 350 mW output power, 1.2 V to 0.6 V output voltage) is taken as a study case. Specifications are added in the landscape and the trend analysis orientates the converter architecture for 40 nm CMOS technology. Particularly, cascode-based power stage is analyzed and selected. Primary post-layout simulation results let hope more than 90% power efficiency. Passive devices are related to an interposer technology targeting the fabrication of magnetic devices on top of embedded capacitor banks.

- Remember PowerSoC 2012
- Inductive vs. capacitive (non-isolated) DC/DC
- Why +100MHz switching frequency DC/DC ?
- A possible lecture of the state-of-the-art
- PowerSWIPE proposal, early results
- Perspectives

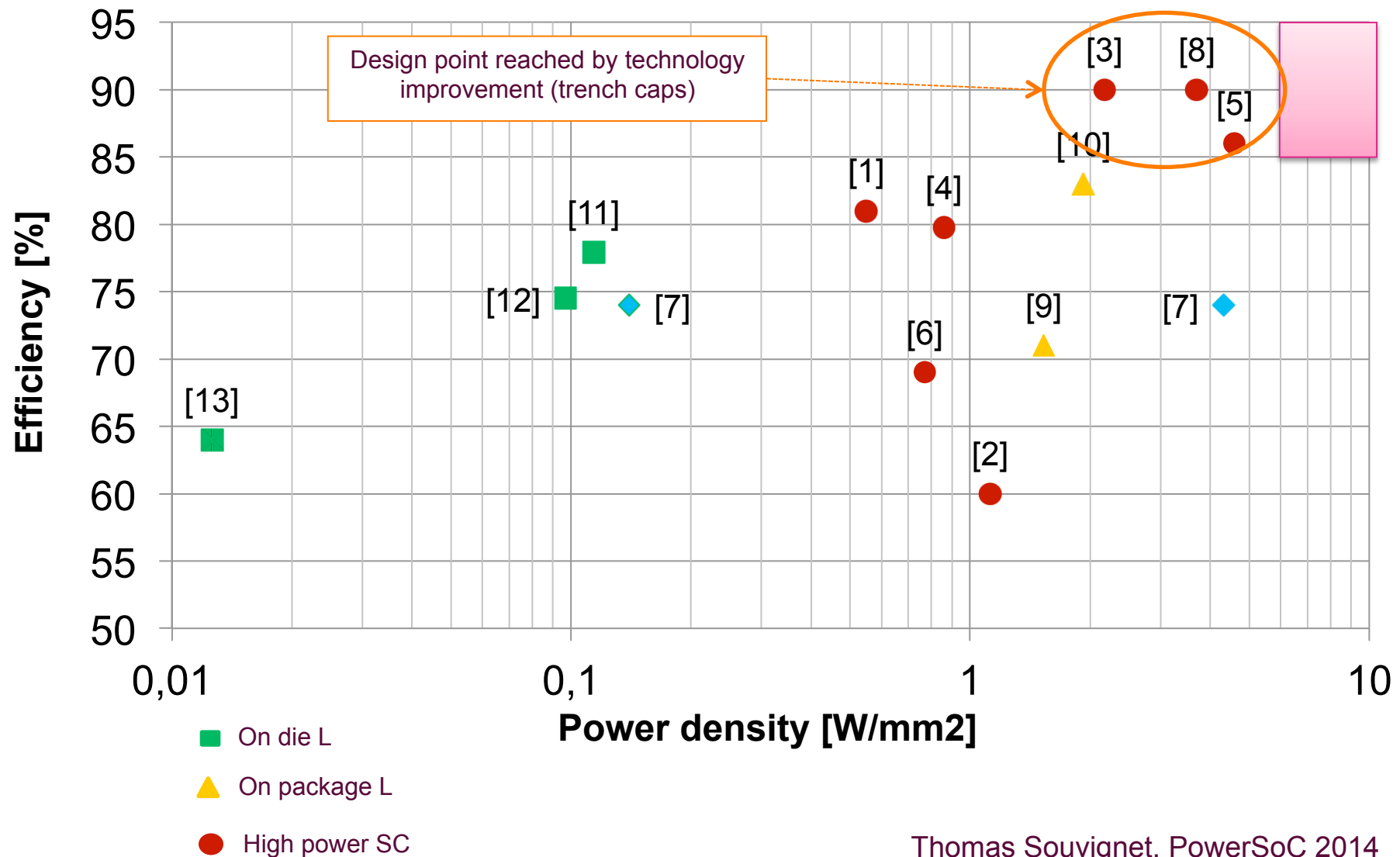
- **Now established: “power is not an afterthought of system”**
- **Specifications now common in essence**
 - how to deliver as much Amps as possible in a given area, for a targeted quality of service, with the best efficiency and the lowest cost
- **Higher expectations: losses < 10% of platform power + efficiency flatness**
- **Global view: passives as important as actives**
- **Complex control schemes despite high switching frequency**

Non-isolated DC/DC converters

↪ Inductive versus capacitive architectures

- SoA → efforts to push the limits
- SoC → heavy technology constraint
- $V_{\text{out}}/V_{\text{in}}$ specification
- Necessity for a portfolio of solutions with various trade-offs

Capacitive on-chip DC/DC



Thomas Souvignet, PowerSoC 2014

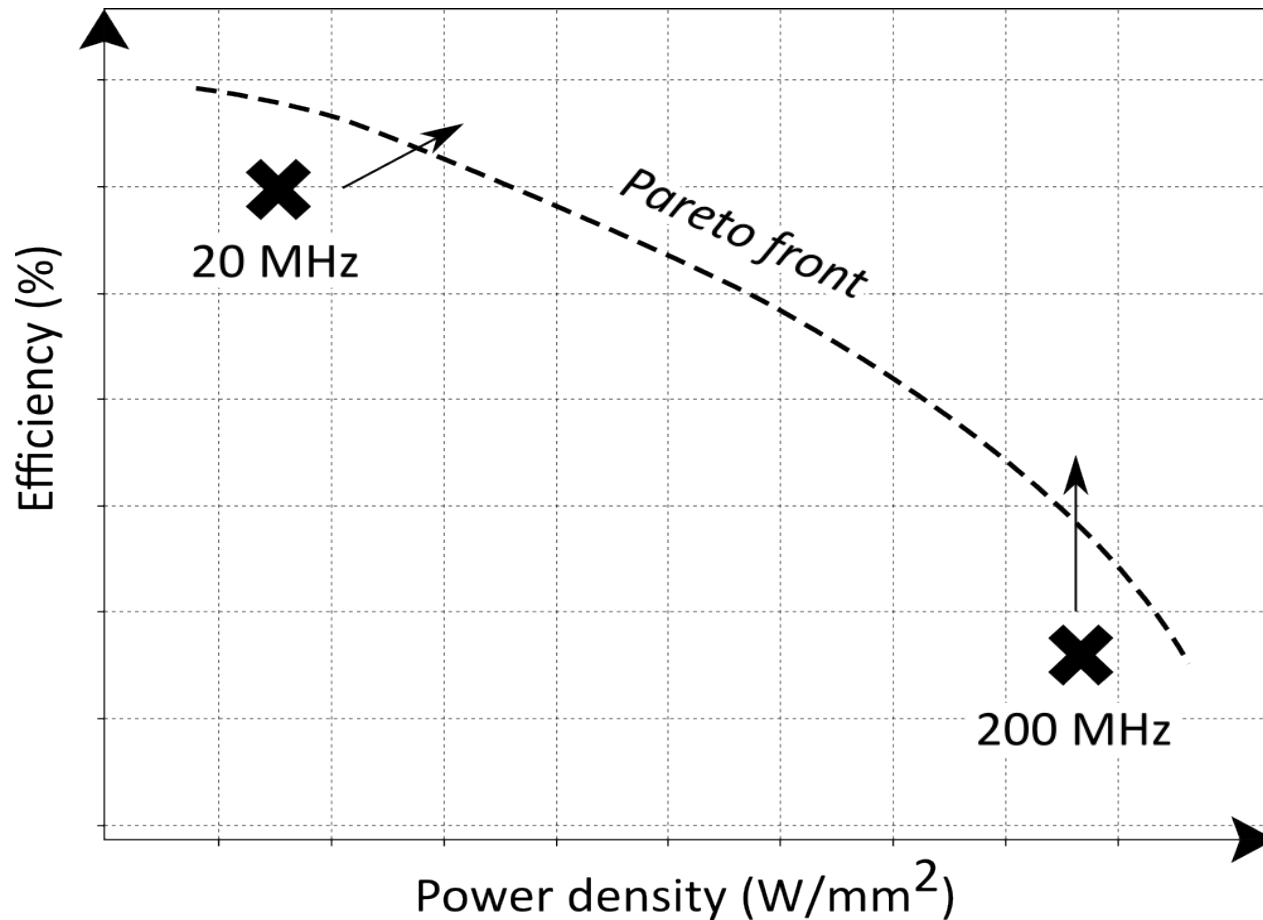
Non-isolated DC/DC converters

➤ Why +100MHz switching frequency ?

- Improvements in technology enables better active and passive devices → mechanical increase in frequency thus in footprint
- Application requirements
 - large transient performances (EER, DVFS)
 - power density (smaller footprint)
- Mission profile (efficiency flatness)

Why +100MHz switching frequency ?

Losses → limit voltage and frequency, but...



J.W. Kolar et al., "PWM Converter Power Density Barriers", IEEE PCC, 2007

↳ Silicon demonstrators

- Steady-state performances
- Transient performances, Power density aspects: no common benchmarks

↳ Many metrics, unique FoM not pertinent → landscapes

↳ Reference point :

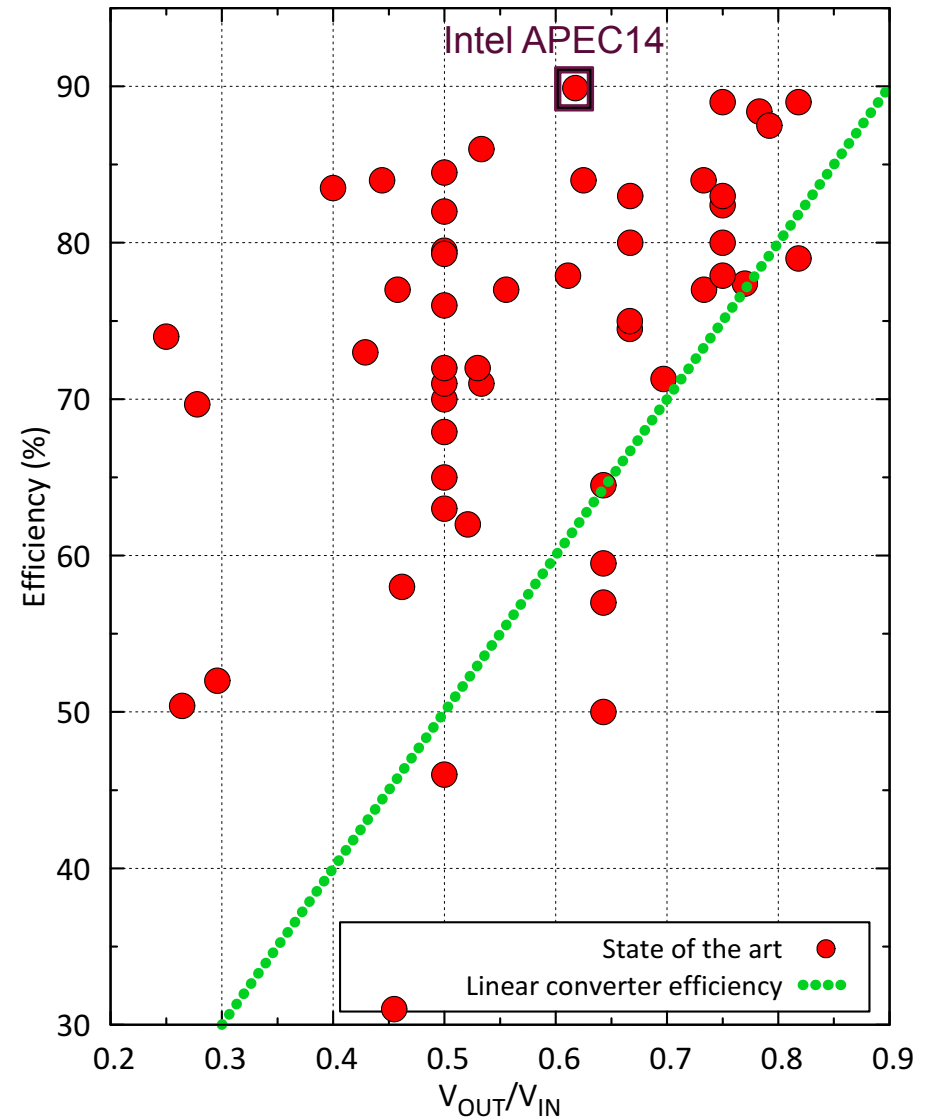
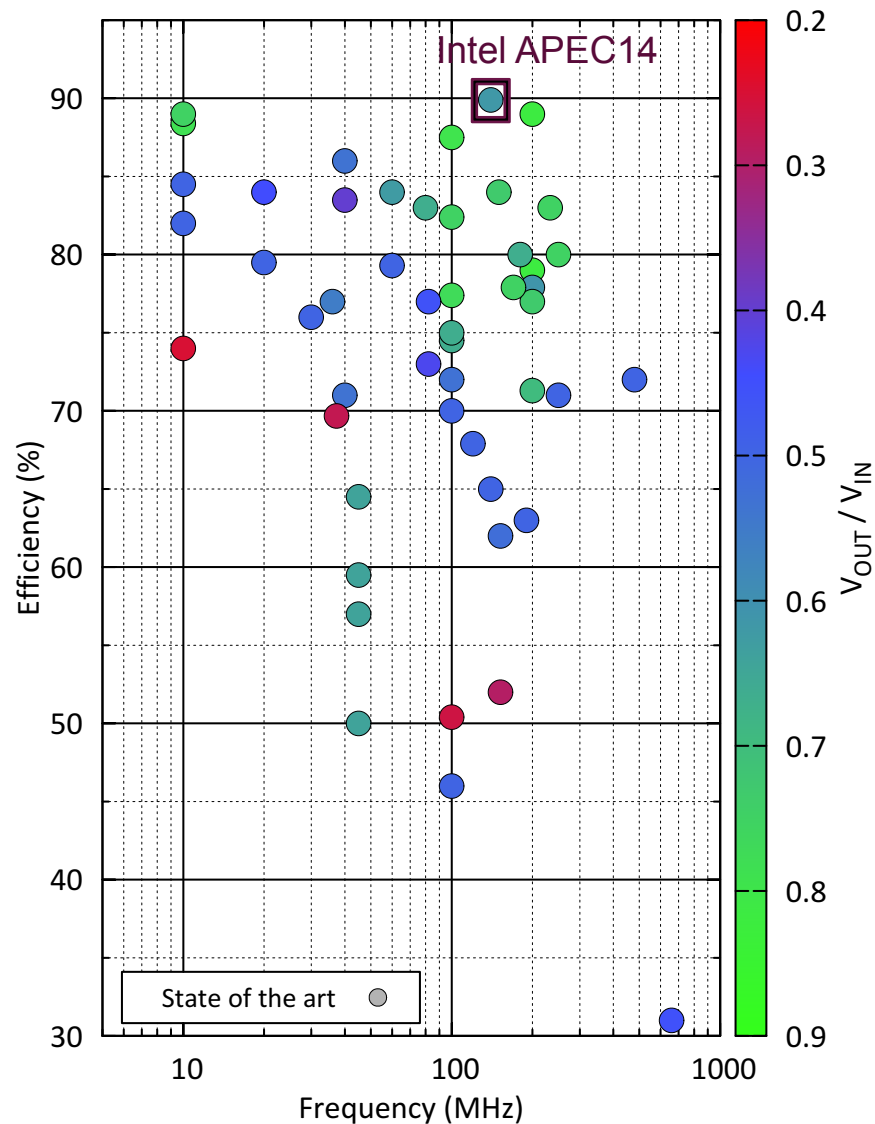
[Burton et al., 2014] Fully integrated voltage regulators on 4th generation Intel core SoCs. In 29th Annual IEEE Applied Power Electronics Conference and Exposition (APEC), 2014, pages 432-439.

CIPS 2014: Neveu, F.; Martin, C.; Allard, B., "Review of high frequency, highly integrated inductive DC-DC converters", 8th International Conference on Integrated Power Systems (CIPS), 2014, pp.1-7, 25-27 Feb. 2014

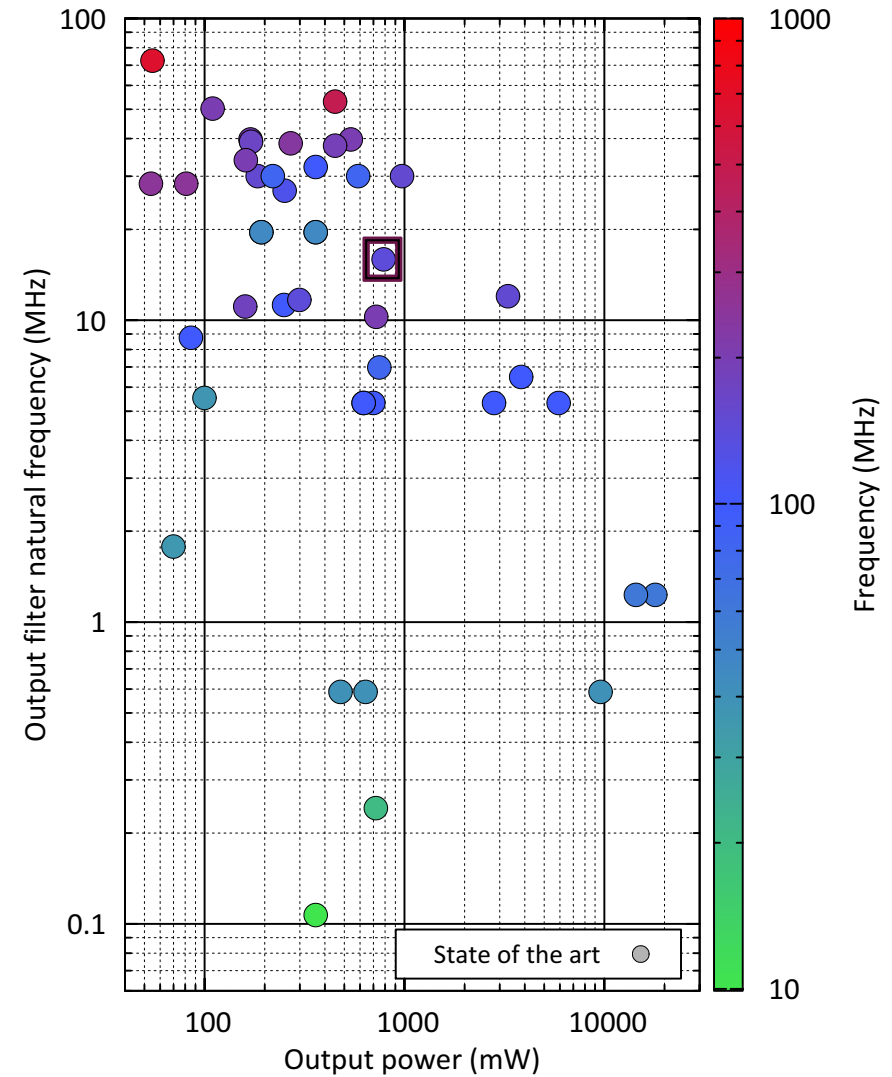
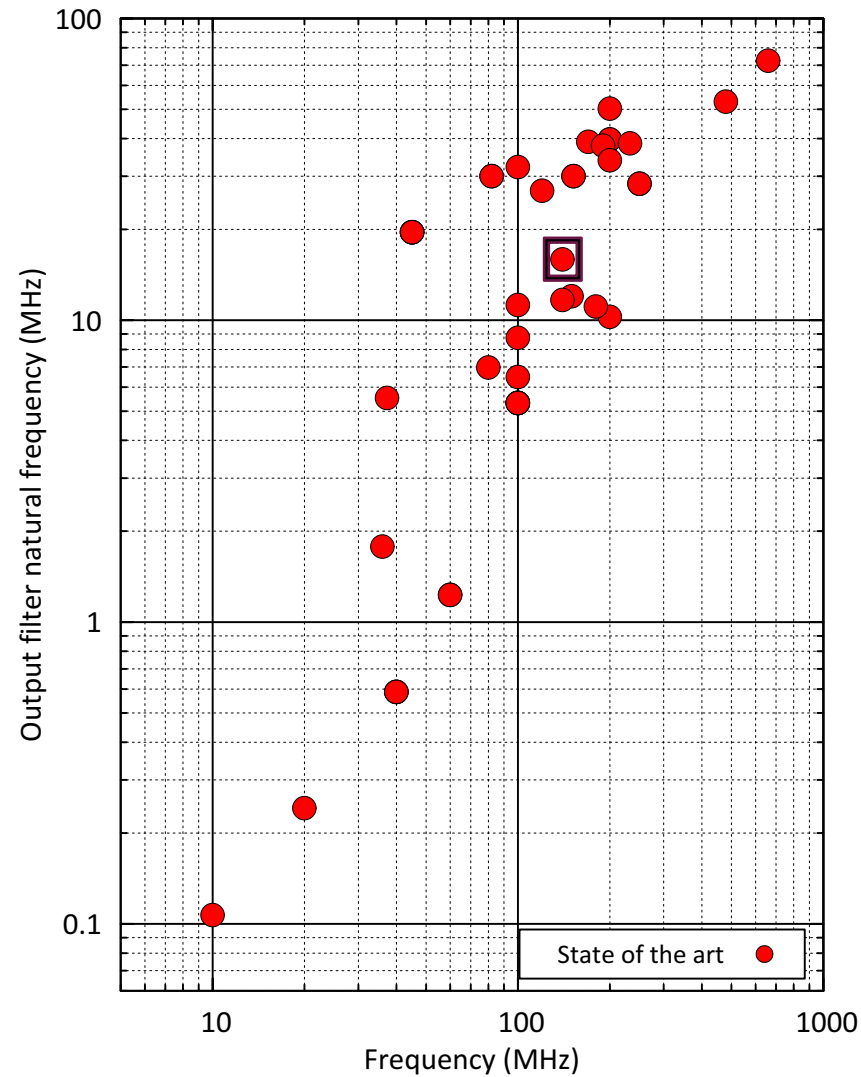
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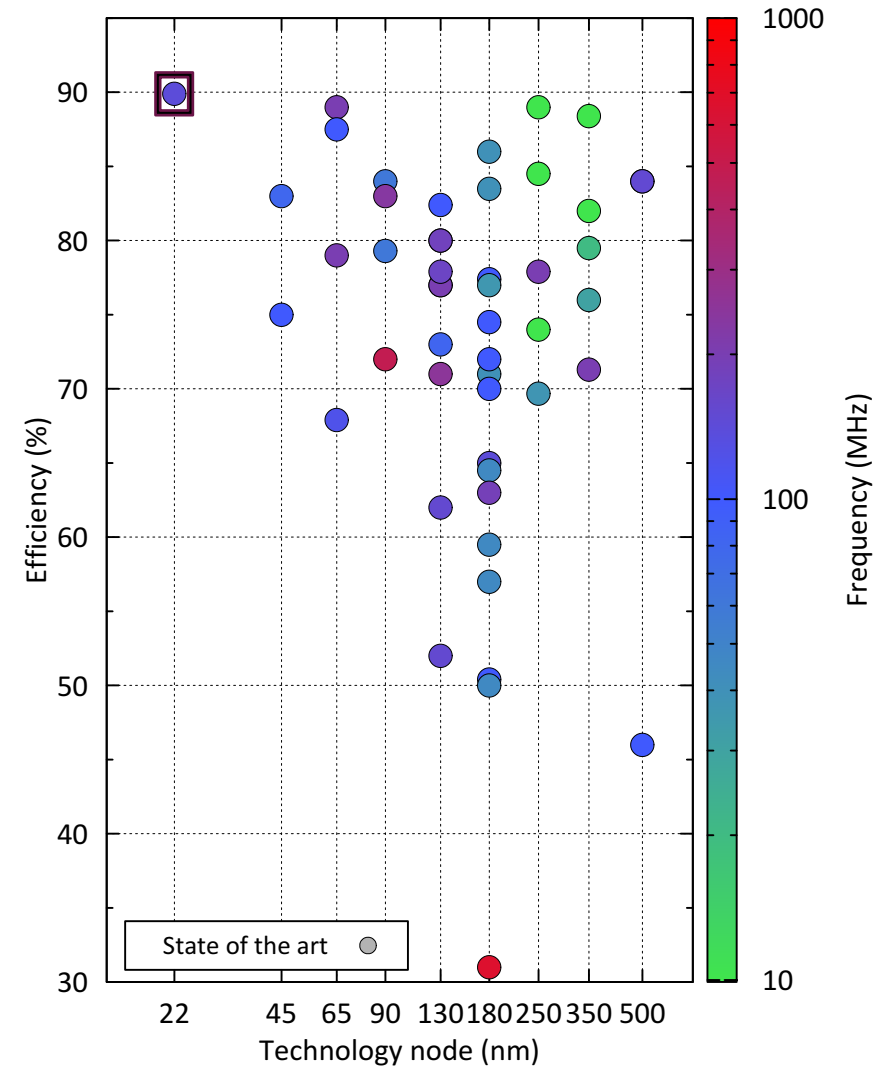
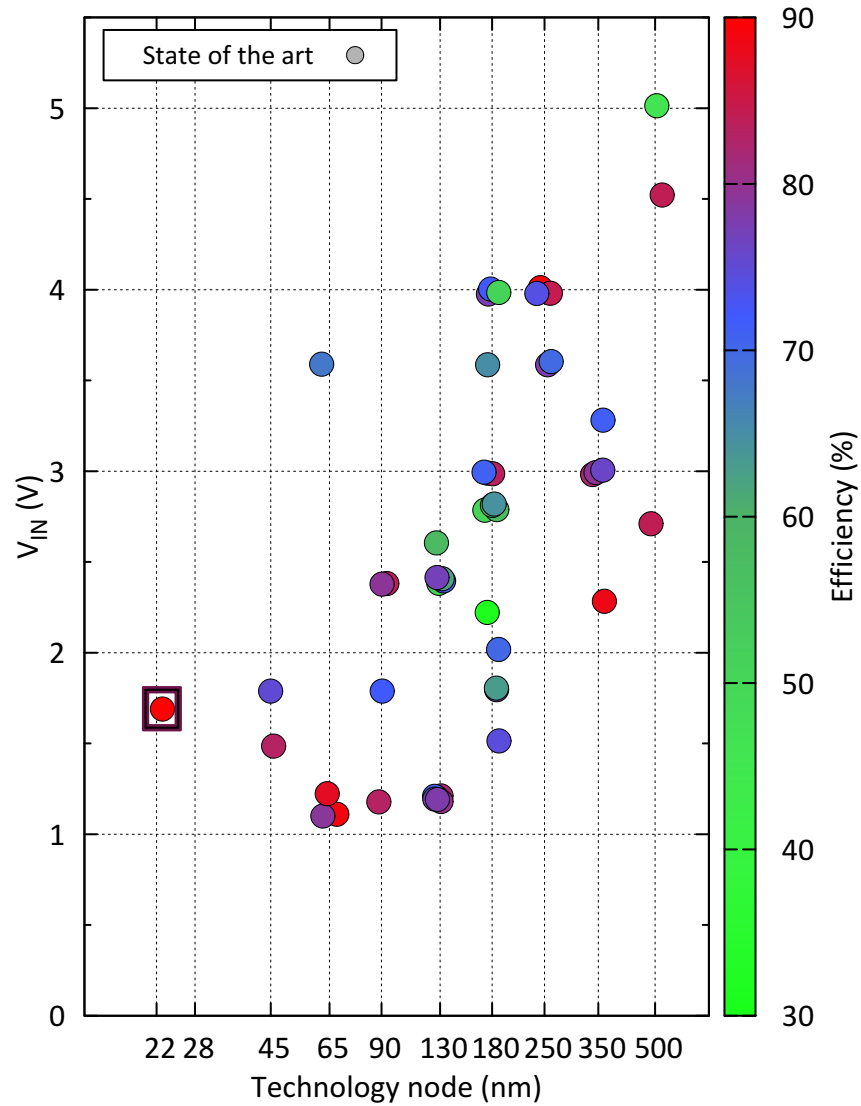
Efficiency, frequency and ratio



Output filter



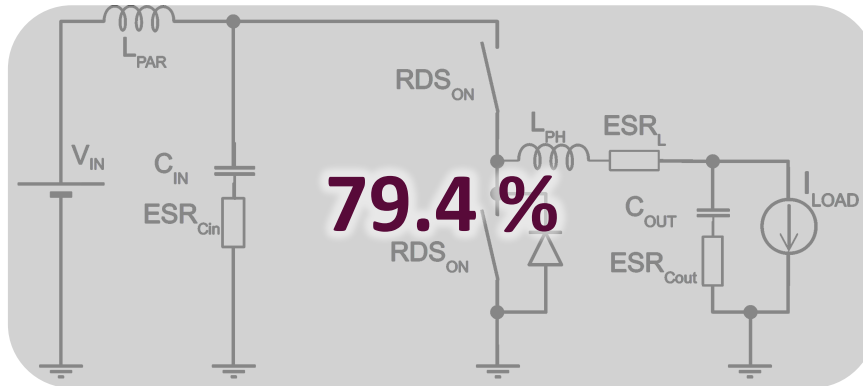
Technology



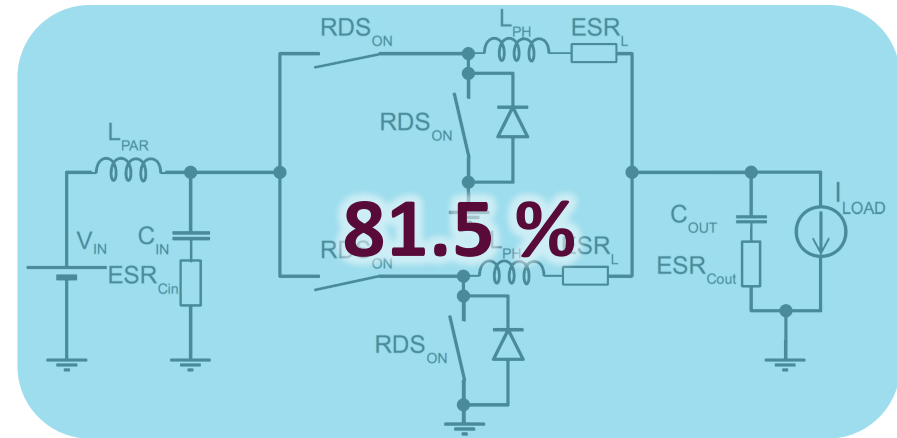
- **Voltage mode versus current mode**
- **PWM versus Hysteretic control or sliding mode control**
- **Digital versus analogue**
 - 2 control loops (analogue control + digital supervisor)
 - Full digital low-losses controller demonstrated (limitations in transient performances)
- **Systematic stability of closed-loop DC/DC quite complex (sample data modeling)**
- **CAE quite complex**

Wrap-up

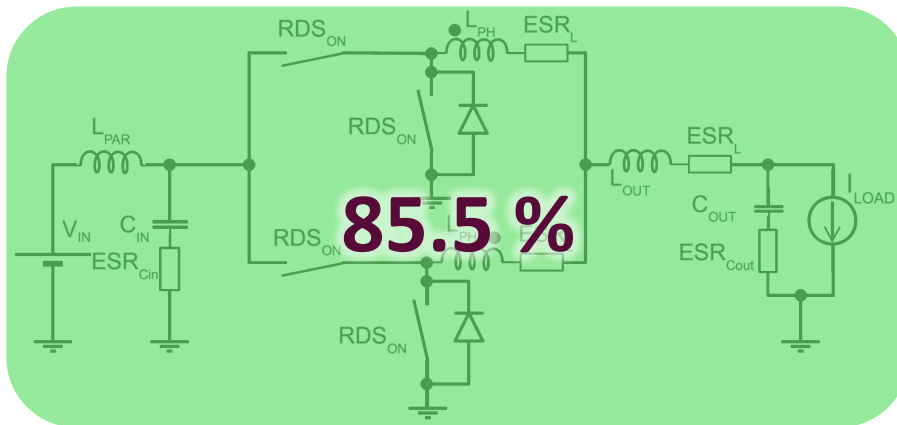
- +100MHz switching frequency → phase coupling



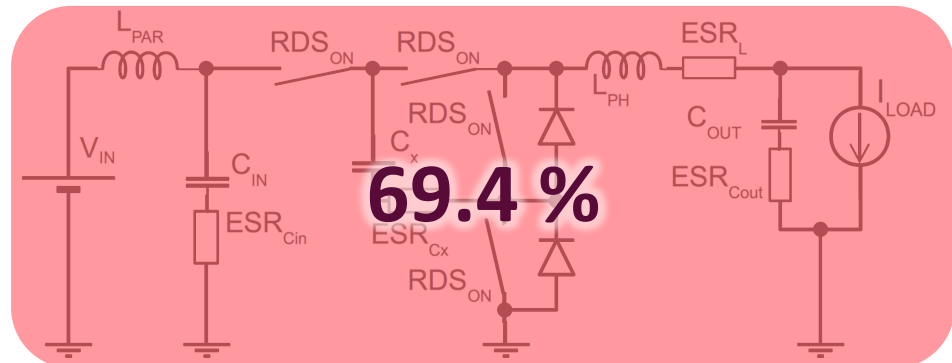
Standard cell



2-phases



Coupled 2-phases



3-levels

↪ **Efficiency flatness → phase shedding**

↪ **Air inductors versus magnetic devices:**

- Cost, complexity
- EMI is an issue
- Coupling of inductors → magnetic material

↪ **Separate issues, separate technologies**

- Interposer of passive devices
- Enable trench capacitors

↪ **Motivations for PowerSWIPE HF DC/DC**

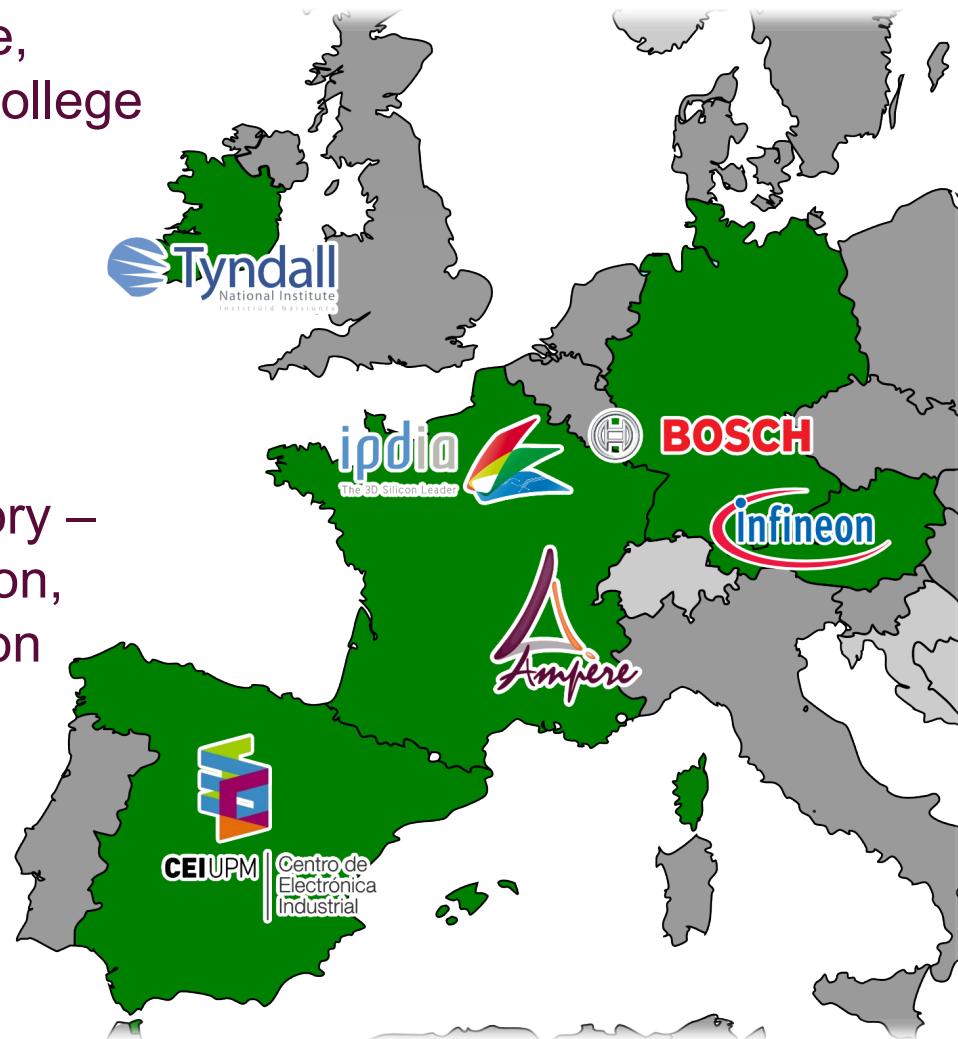
PowerSWIPE consortium

Tyndall National
Institute,
University College
Cork

IPDiA,
Caen

Ampère Laboratory –
University of Lyon,
INSA Lyon, Lyon

Centro de Electrónica
Industrial –
Universidad
Politécnica de Madrid



Robert Bosch
GmbH

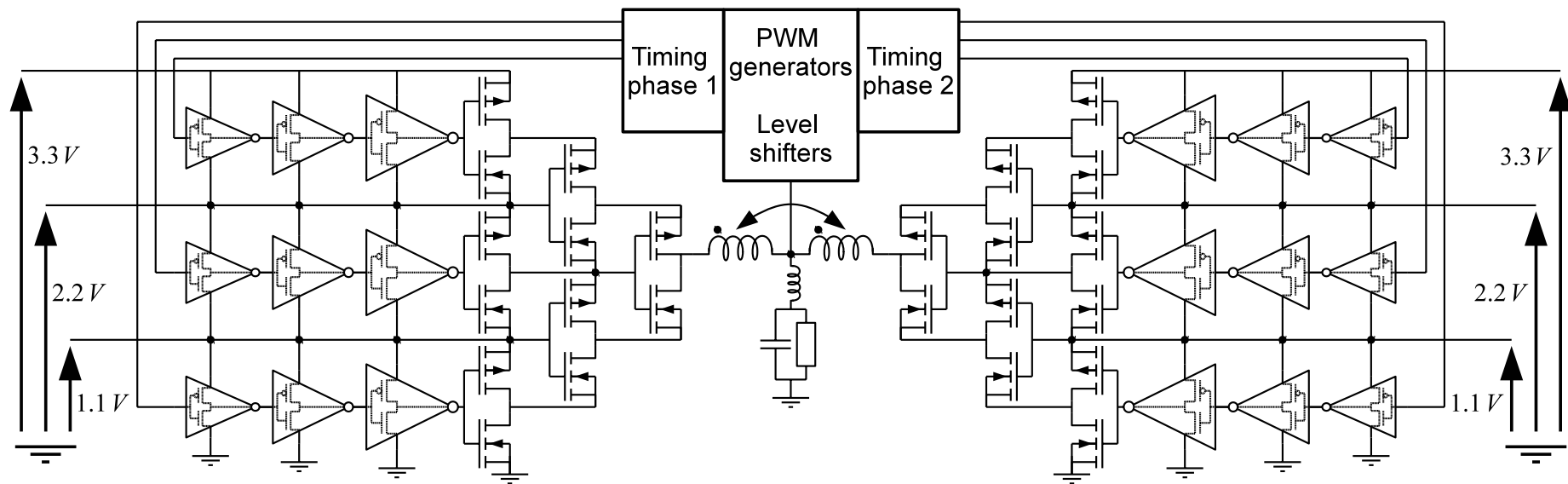
Infineon
Technologies
AG

Infineon
Technologies
Austria Villach
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Kursun, V., Narendra, S. G., De, V. K., & Friedman, E. G. (2005). Cascode monolithic DC-DC converter for reliable operation at high input voltages. *Analog Integrated Circuits and Signal Processing*, 42(3), 231-238 → **evaluation by simulation**

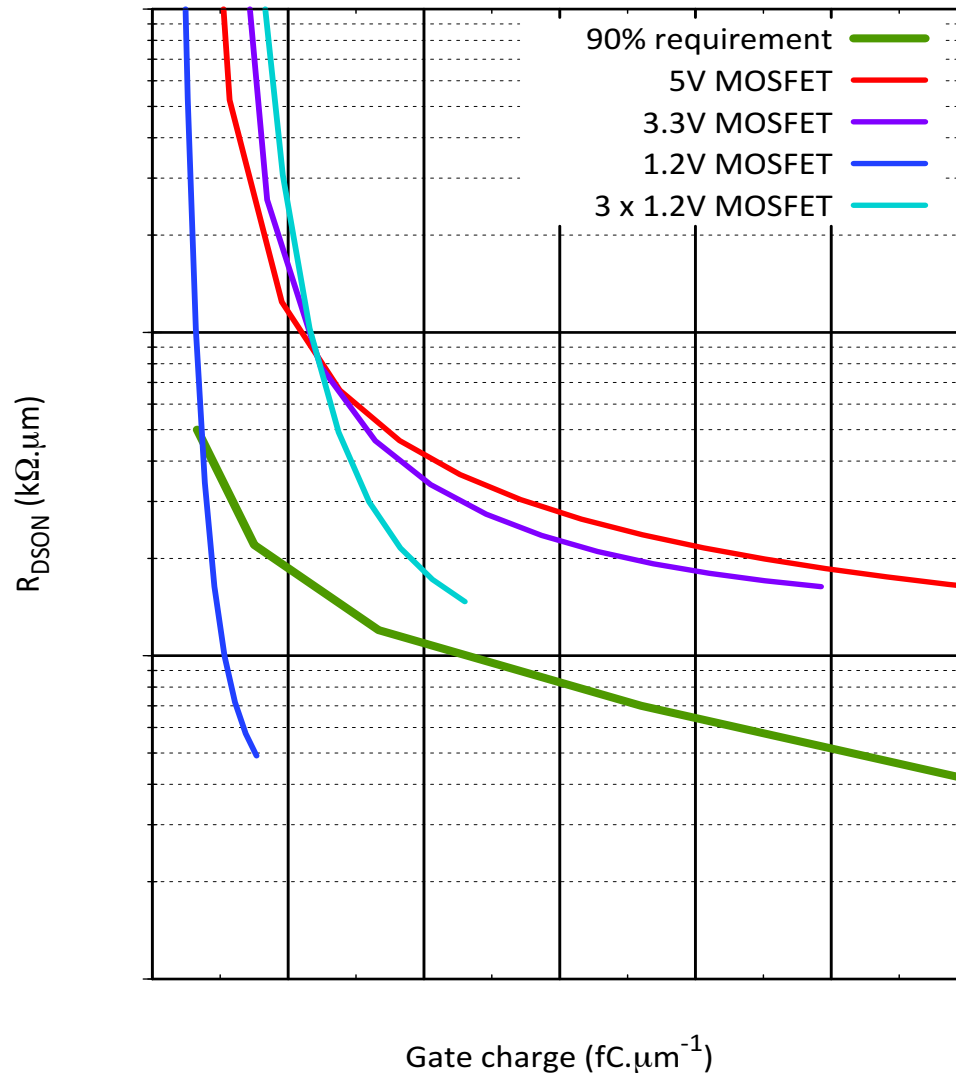
Peng et al., 2013, A 100 MHz two-phase four-segment DC-DC converter with light load efficiency enhancement in 0.18 μm CMOS. *IEEE Trans. Circ. & Syst. I*, 60(8):2213–2224.

- ↪ **40nm bulk CMOS**
- ↪ **Interposer by IPDIA and Tyndall**
- ↪ **Hard switching then Hysteretic control**



See details on design on poster #9 by Florian Neveu

Motivation for low-voltage MOSFET



➡ **Standard DC/DC step-down**

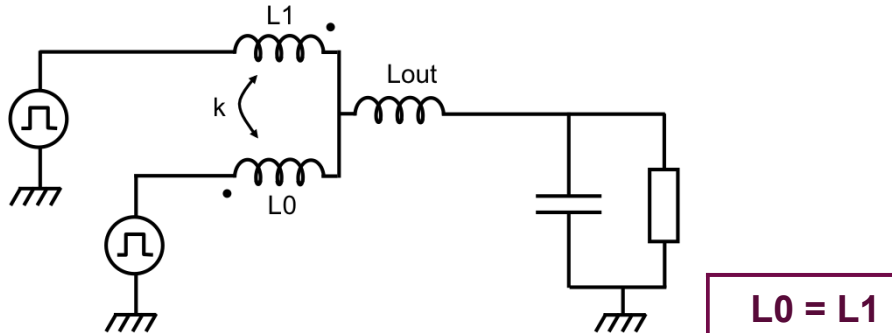
➡ **Analytical model of losses**

➡ **Low-voltage MOSFET**

➡ **Digital MOSFET could be considered**

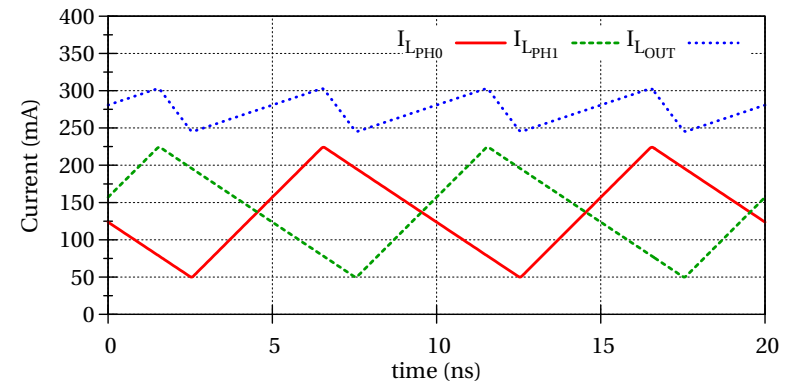
➡ **Similar conclusions in any thin technology**

Local optimization of coupled inductors

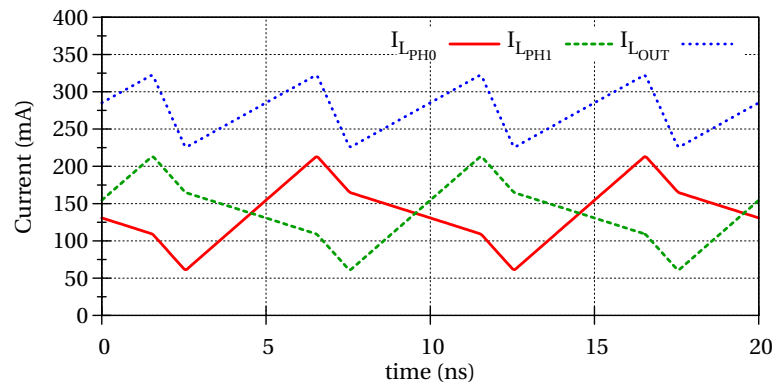



Minimizing phase current ripple

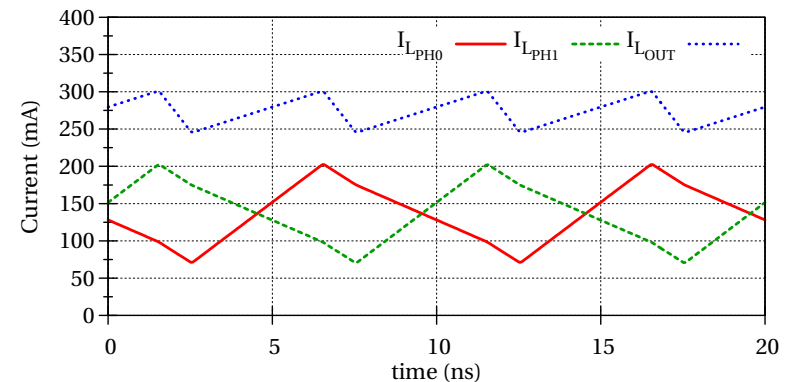
$$\begin{aligned}
 &L_0 = L_1 \\
 &-1 < k < 1 \\
 &0 \leq L_{out} \leq 91 \text{ nH} \\
 &L_0 + L_1 + L_{out} = 91 \text{ nH}
 \end{aligned}$$



#1: $k = 0$, $L_0 = L_1 = 45.5 \text{ nH}$, $L_{out} = 0 \text{ nH}$
Ripple: Phases: 172 mA, Output: 56 mA

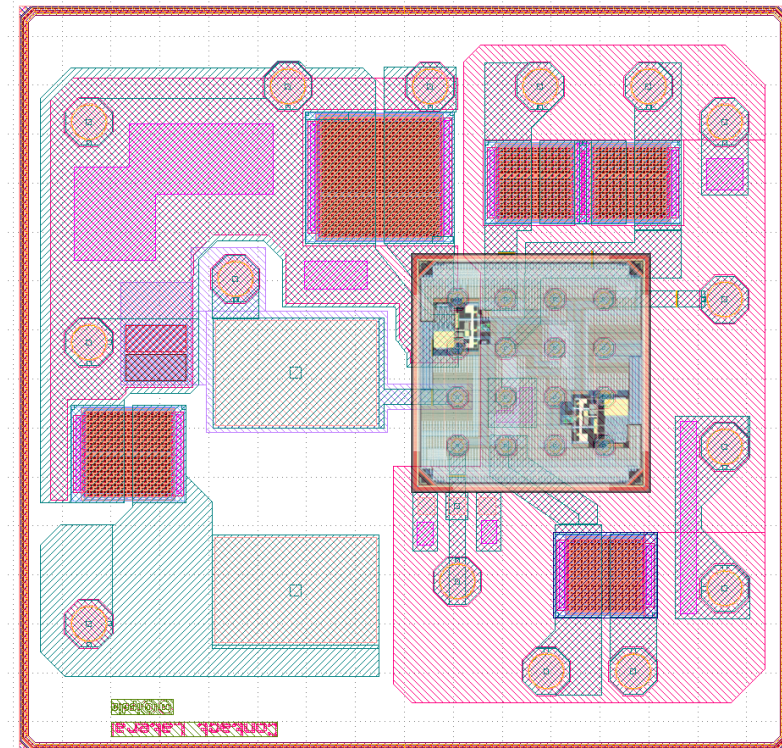
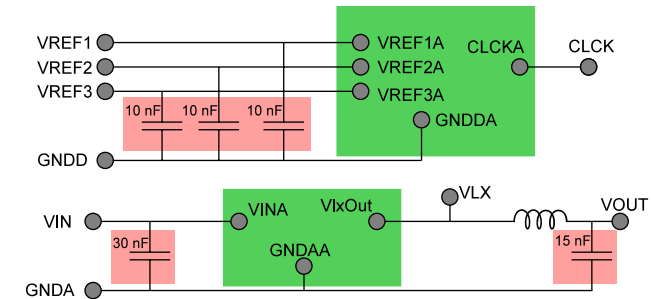
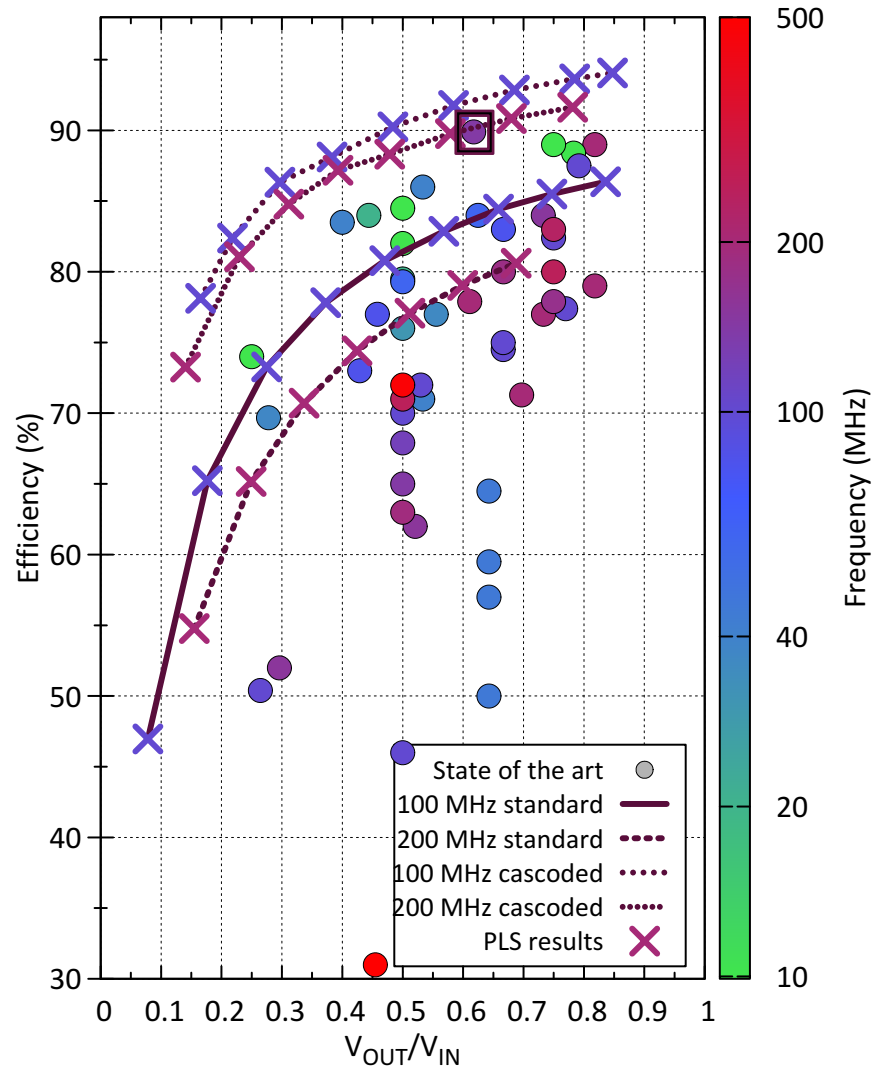


#2 : $L_{out} = 0 \text{ nH}$, $L_0 = L_1 = 45.5 \text{ nH}$, $k = 0.4$
Ripple: Phases: 151 mA, Output: 102 mA



#3: $L_0 = L_1 = 35 \text{ nH}$, $L_{out} = 21 \text{ nH}$, $k = 1$
Ripple: Phases: 123 mA, Output: 55 mA

Primary PLS results



- **Waiting for silicon delivery**
- **Multi-chip test board**
- **Verification of limits of robustness**

- **Multi-phase design**
- **Optimized IC layout**
- **Optimized interposer layout**